

Non-Parametric Signal Detection and RMT

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Talk Outline

1. Signal Detection Problems
2. Connection to Random Matrix Theory
3. Implications

Detection of signals embedded in noise

Given a measurement system with p sensors (antennas / microphones / hyperspectral camera / etc)

Observe multivariate samples $\mathbf{x}(t_j) \in \mathbb{R}^p$ of the form

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where

- ▶ $\mathbf{s}(t) = (s_1(t), \dots, s_K(t))'$ are K time-dependent signals.
- ▶ \mathbf{A} is a $p \times K$ *fixed* unknown mixing matrix of rank K (steering matrix).

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- ▶ \mathbf{A} is a $p \times K$ *fixed* unknown mixing matrix of rank K (steering matrix).
- ▶ σ - noise level, $\xi \sim \mathcal{N}(0, \mathbf{I}_p)$.

Problem Formulation

$$\mathbf{x}(t) = \mathbf{A}\mathbf{s}(t) + \sigma\xi(t) \quad (*)$$

Assume $\mathbf{s}(t) \in \mathbb{R}^K$ is stationary random process with a full rank covariance matrix, and that mixing matrix \mathbf{A} is of rank K (e.g., there are indeed K identifiable sources).

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Given n i.i.d. samples \mathbf{x}_i from the model (*), estimate the number of sources K .

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The linear mixture (factor) model

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- ▶ Analytical Chemistry / Chemometrics: \mathbf{x} is the measured (logarithm of) spectra at p wavelengths, \mathbf{s} - vector of concentrations of K chemical components. Eq. (*) follows from Beer-Lambert's law.
- ▶ Signal Processing: \mathbf{s} is a vector of K emitting sources, \mathbf{x} - measurement at an array of p receivers (microphones, antennas, etc).

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- ▶ Signal Processing: \mathbf{s} is a vector of K emitting sources, \mathbf{x} - measurement at an array of p receivers (microphones, antennas, etc).
- ▶ Statistical Modeling at large : linear mixture models / factor models / error in variables / two-way table with multiplicative interactions.

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In signal processing -

a preliminary step before blind source separation / independent component analysis, direction of arrival estimation, many other parametric procedures whose number of parameters depends on K - the number of sources.

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In chemometrics -

process control, optimal number of latent variables in regression/calibration models, first step prior to self modeling curve resolution and many other estimation procedures.

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In this setting, assuming Gaussian signals, eigenvalues $\ell_1 \geq \ell_2 \geq \dots \geq \ell_p$ of the sample covariance matrix

$$\mathbf{S}_n = \sum_i \mathbf{x}_i' \mathbf{x}_i$$

are *sufficient statistics* for eigenvalues of the population covariance matrix $\mathbf{\Sigma}$ [James 66', Muirhead 78'].

Detection of Signals as a model selection problem

Given K sources, the population covariance matrix of the observations has diagonal form

$$\mathbf{W}'\Sigma\mathbf{W} = \sigma^2\mathbf{I}_p + \text{diag}(\lambda_1, \lambda_2, \dots, \lambda_K, 0, \dots, 0) \quad (1)$$

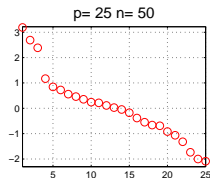
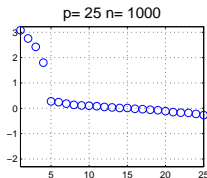
Model Selection Problem: Given $\{\ell_j\}_{j=1}^p$, determine which model of the form (1) is most likely.

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Eigenvalue Based Rank Estimation

Challenge: Distinguish between large yet insignificant noise eigenvalues and small yet significant signal eigenvalues.

Problem Parameters: n - number of samples, p - dimensionality, σ - noise level.

In classical array processing $p/n \ll 1$,

In chemometrics and in some modern radar/sonar systems:
 $p/n = O(1)$ and often $p/n \gg 1$.

Previous Approaches

Can be divided into 2 main disciplines,

- ▶ Nested Hypothesis Tests (with various test statistics)
- ▶ Information Theoretic Criteria (BIC, MDL, AIC, etc)

Previous Approaches

The "godfathers"

- Bartlett (1940's), Lawley (1950's) - likelihood ratio tests, tests for sphericity, assume Gaussian observations, asymptotic expansions for p fixed, $n \rightarrow \infty$.

$$T = \frac{(\prod \ell_i)^{1/p}}{\frac{1}{p} \sum \ell_i}$$

This statistic does not work well when p is of the same order of n and is undefined if $p > n$.

Previous Approaches

In chemometrics ($p \gg n$):

- Malinowski F-test, 1977, 1980, 1987-1990, *Analytical Chemistry, J. Chemometrics*.
- **Faber and Kowalski, Modification of Malinowski's F-test, J. Chemometrics, 1997.**
- Faber, Buydens and Kateman, Aspects of pseudorank estimation methods based on the eigenvalues of principal component analysis of random matrices, 1994.
- at least 15 other papers describing and comparing various algorithms.

Previous Approaches

In signal processing and statistics literature:

- Wax & Kailath, *Detection of signals by information theoretic criteria*, 85'
- Zhao, Krishnaiah and Bai. JMVA, 1986
- Fishler, Grossmann and Messer, IEEE Sig. Proc. 2002.
- P-J. Chung, J.F. Böhme, C.F. Mecklenbraüker and A.O. Hero, 2007.
- many other papers in signal processing and in statistics.
- Silverstein & Combettes, 1992
- Schott, *A high-dimensional test for the equality of the smallest eigenvalues of a covariance matrix*, JMVA, 2006.
- Rao & Edelman, *Sample eigenvalue based detection of high-dimensional signals in white noise using relatively few samples*. 2007.

Detection of Signals by Information Theoretic Criteria

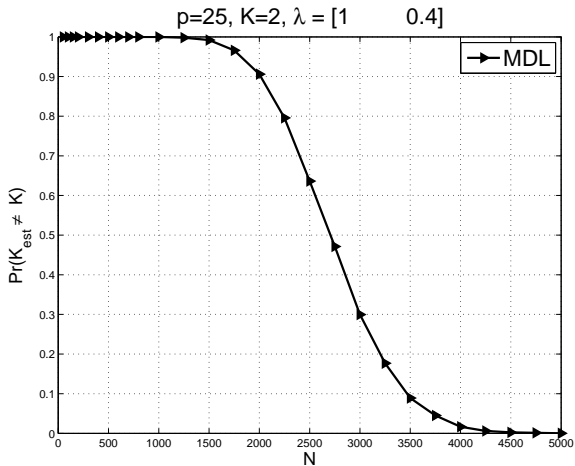
$$\hat{k}_{MDL} = \arg \min \quad -(p-k)n \left[\frac{1}{p-k} \log \left(\prod_{k+1}^p \ell_j \right) - \log \frac{\sum_{k+1}^p \ell_j}{p-k} \right] \\ + \frac{1}{2}k(2p-k) \log n$$

$$\hat{k}_{AIC} = \arg \min \quad -(p-k)n \left[\frac{1}{p-k} \log \left(\prod_{k+1}^p \ell_j \right) - \log \frac{\sum_{k+1}^p \ell_j}{p-k} \right] \\ + k(2p-k)$$

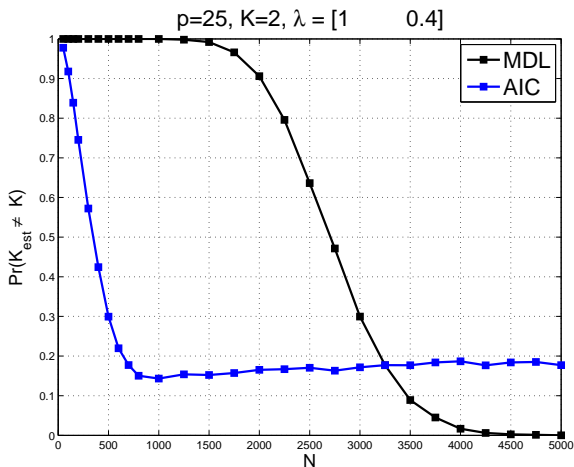
MDL estimator became the standard tool in signal processing.

WAX & KAILATH 1985.

Example: MDL consistent



Example: AIC not consistent



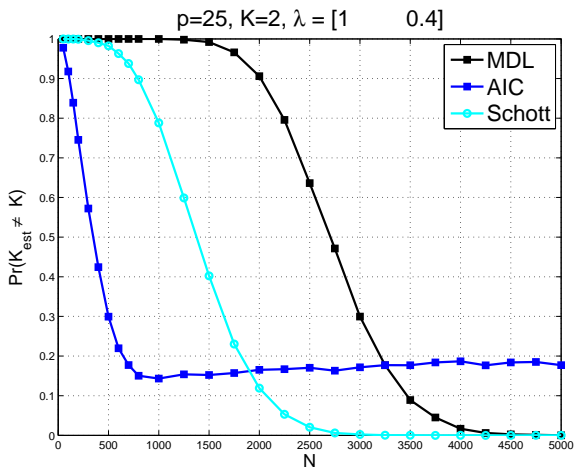
Some Modern Approaches:

Rao & Edelman (2007) and Schott (2006)

$$T = \frac{\sum \ell_j^2}{(\sum \ell_j)^2}$$

This statistic can be used for sphericity test for all values of p/n (Ledoit & Wolf).

Example:



Detection of Signals - Theoretical Questions

- ▶ Which Test Statistic to use ?
- ▶ Why does AIC overestimate number of signals ?
- ▶ Detection Performance ?
- ▶ Known vs. unknown noise level
- ▶ Non-parametric vs fully parametric setting.

KEY TOOLS

- ▶ Behavior of noise eigenvalues.

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TOOLS:

- RMT
- Matrix Perturbation Theory

Which Test Statistic ?

Consider the case of two (nearly) *simple* hypothesis

$$\mathcal{H}_0 : \mathbf{\Sigma} = \mathbf{I} \quad \text{vs.} \quad \mathcal{H}_1 : \mathbf{W}'\mathbf{\Sigma}\mathbf{W} = \mathbf{I} + \text{diag}(\lambda, 0, \dots, 0)$$

with λ - known. What is unknown is the basis which makes $\mathbf{\Sigma}$ diagonal in \mathcal{H}_1 .

Neyman-Pearson: optimal method is likelihood ratio test

$$\frac{p(l_1, \dots, l_p | \mathcal{H}_1)}{p(l_1, \dots, l_p | \mathcal{H}_0)} \geq C(\alpha)$$

Which Test Statistic ?

From multivariate analysis (Muirhead 78')

$$p(\ell_1, \dots, \ell_p | \Sigma) = C_{n,p} \prod_i \ell_i^{(n-p-1)/2} \prod_{i < j} (\ell_i - \ell_j) {}_0F_0\left(-\frac{1}{2}nL, \Sigma^{-1}\right)$$

${}_0F_0$ - hypergeometric function with matrix argument.

Key point: asymptotically in n , for p fixed,

$$\log \left(\frac{p(\ell_1, \dots, \ell_p | \mathcal{H}_1)}{p(\ell_1, \dots, \ell_p | \mathcal{H}_0)} \right) \approx n(\ell_1 - h(\lambda)) + O\left(\sum c_{1j}/(\ell_1 - \ell_j)\right)$$

Asymptotically, should only look at *largest* eigenvalue !

Roy's Largest Eigenvalue Test

If $\ell_1 > \sigma^2 \cdot th(\alpha)$ accept \mathcal{H}_1 – signal is present.

$th(\alpha)$ - found by distribution of ℓ_1 under the null of no signals.
Then, $S_n = 1/n \sum_i \mathbf{x}_i \mathbf{x}_i'$ is a scaled Wishart matrix.

Theorem: [Johansson 00', Johnstone 01', El-Karoui 07'] As
 $p, n \rightarrow \infty$

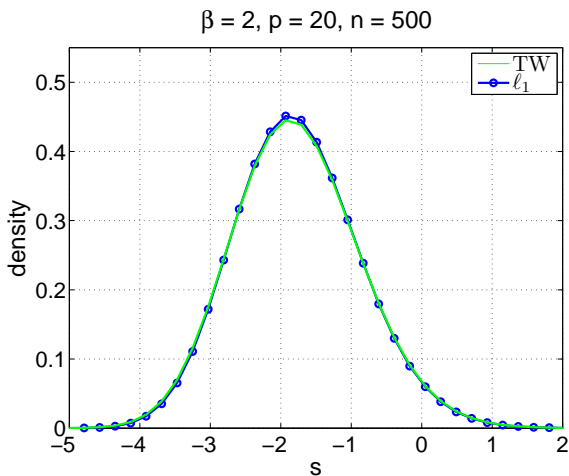
$$\Pr\{\ell_1 < \sigma^2 (\mu_{n,p} + s\sigma_{n,p})\} \rightarrow TW_{\beta}(s)$$

where TW_{β} - Tracy-Widom distribution of order β

$\beta = 1$ - real valued noise, $\beta = 2$ - complex valued noise.

For any confidence level α can invert TW distribution to obtain threshold $s(\alpha)$.

Largest Eigenvalue Distribution



Unknown Noise Level

Standard approach: replace σ^2 by its ML estimate

$$\hat{\sigma}^2 = \frac{1}{p} \sum_{j=1}^p \ell_j$$

GLRT:

$$U = \frac{\ell_1}{\frac{1}{p} \sum_j \ell_j} > th(\alpha)$$

This test statistic plays a role in:

- ▶ Signal Detection [Besson & Scharf 06', Kritchman & N. 08, Bianchi et al. 09']
- ▶ Two-way models of interaction [Johnson & Graybill, 72']
- ▶ Models for Quantum Information Channels.

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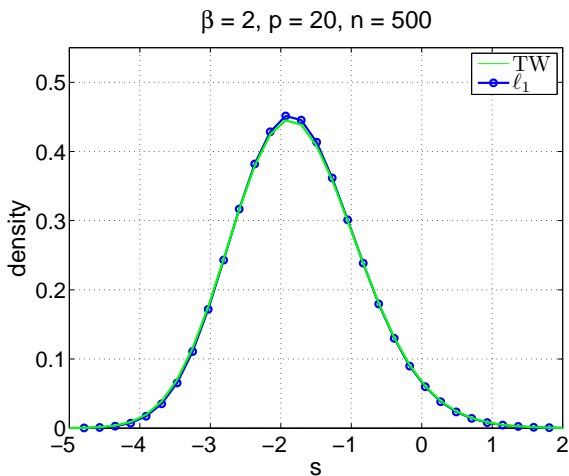
Unknown Noise Level

In principle, can use same threshold $th(\alpha)$, since:

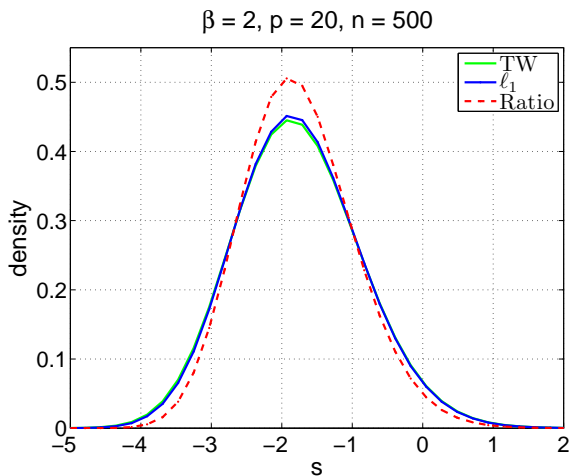
Theorem:[Bianchi et. al.] As $p, n \rightarrow \infty$

$$\Pr[U < \mu_{n,p} + s\sigma_{n,p}] \rightarrow TW_{\beta}(s)$$

Ratio Distribution



Ratio Distribution



Ratio Distribution

Definition:

$$F_{np}(s) = \Pr \left[\frac{\ell_1 - \mu_{np}}{\sigma_{np}} < s \right], \quad H_{np}(s) = \Pr \left[\frac{U - \mu_{np}}{\sigma_{np}} < s \right].$$

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Key property: U and $T = \frac{1}{p} \sum_j \ell_j$ are independent.

Ratio Distribution

Since $\ell_1 = U \cdot T$,

$$\Pr[\ell_1 < x] = \int_0^{px} \Pr\left[U < \frac{x}{t}\right] p_T(t) dt$$

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L.H.S. = approximately TW

R.H.S. = convolution of required function with χ^2 density.

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(i) uniformly in p and s , $H_{np}(s)$ is a smooth function with bounded third derivative, $|H_{np}'''(s)| < C$.

(ii)

$$|F'_{np}(s) - TW'_{\beta}(s)| \rightarrow 0 \quad \text{and} \quad |F''_{np}(s) - TW''_{\beta}(s)| \rightarrow 0 \quad (2)$$

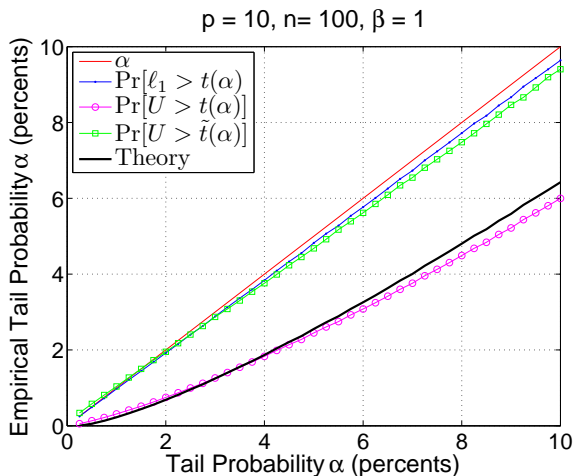
Ratio Distribution

Theorem: In the joint limit $p, n \rightarrow \infty$,

$$H_{np}(s) - TW_{\beta}(s) = [F_{np}(s) - TW_{\beta}(s)] - \frac{1}{2} \left(\frac{2}{\beta np} \right) \left(\frac{\mu_{np}}{\sigma_{np}} \right)^2 TW_{\beta}''(s) + o(p^{-2/3}).$$

[N., to appear in JMVA]

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Example: $n \gg p$, $s = -0.2325$, where

$$1 - TW_2(s) \approx 5\%$$

Then

$$\frac{|TW_2''(s)|}{1 - TW_2(s)} \approx 7 \quad \text{and} \quad \frac{1}{np} \left(\frac{\mu_{np}}{\sigma_{np}} \right)^2 \approx 1/p^{2/3}.$$

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Hence, for a 10% relative error,

$$\frac{1}{2} \frac{2}{\beta np} \left(\frac{\mu_{np}}{\sigma_{np}} \right)^2 \frac{|TW_2''(s)|}{1 - TW_2(s)} \leq 0.1$$

we need $p \gtrsim (35)^{3/2} \approx 200$.

Detection of Signals - Theoretical Questions

- ▶ Which Test Statistic to use ?
- ▶ **Why does AIC overestimate number of signals**
- ▶ Detection Performance ?
- ▶ Known vs. unknown noise level
- ▶ Non-parametric vs fully parametric setting.

Overestimation Probability of AIC (no signals)

AIC overestimates number of signals when

$$-\ln \mathcal{L}_0 > -\ln \mathcal{L}_1 + \frac{2p-1}{n}$$

where

$$\mathcal{L}_k = \frac{\prod_{i>k} \ell_i}{\left(\frac{\sum_{i>k} \ell_i}{p-k}\right)}$$

Overestimation Probability of AIC

Let ξ be solution of

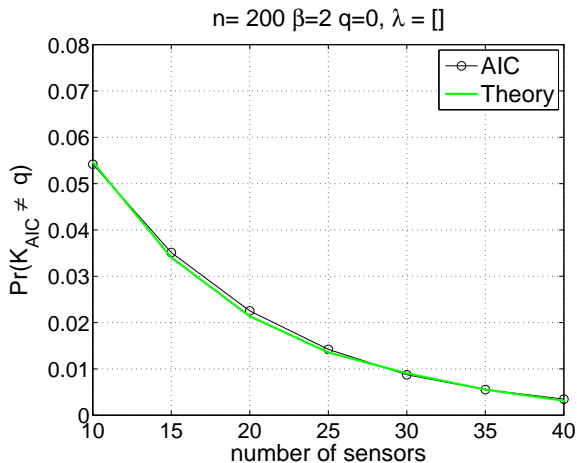
$$-\ln(1 + \xi) - (p - 1) \ln\left(1 - \frac{\xi}{p - 1}\right) = \frac{2p - 1}{n}$$

Lemma:

$$\Pr[k_{\text{AIC}} > 0] = \Pr[U > 1 + \xi] + O\left(\frac{1}{n}\right)$$

[N., IEEE-TSP 10']

AIC Overestimation Probability



AIC Overestimation Probability

Since

$$\xi = 2\sqrt{\frac{p-1/2}{n}}\sqrt{1-\frac{1}{p}}\left(1+O\left(\frac{1}{n^{1/2}}\right)\right)$$

AIC penalty is not sufficiently strong.

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Legacy System:

$$\hat{k}_{AIC} = \arg \min_k \quad -\ln \mathcal{L}_k + \frac{2Ck}{n} \left(p + 1 - \frac{k+1}{2} \right)$$

If $C = 1$ penalty comparable to original AIC.

Modified AIC

Theorem: Modified AIC estimator with $C = 2$ has negligible overestimation probability, which for large n is exponential small in p .

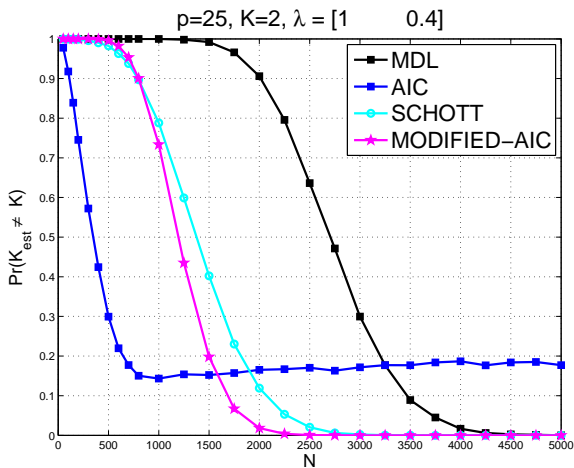
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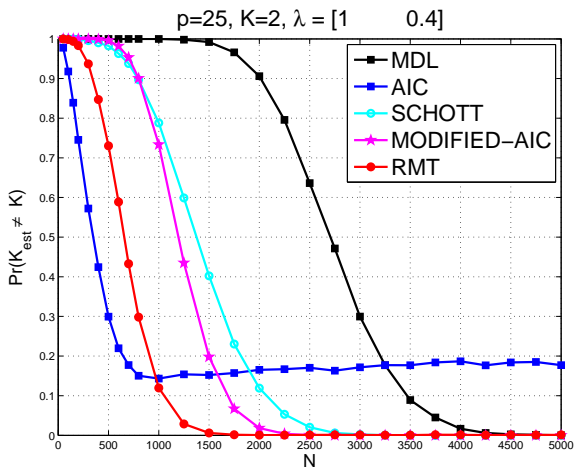
Basic idea: use non-asymptotic bound of Ledoux, valid for all p, n (complex)

$$\Pr[\ell_1 > (1 + \sqrt{p/n})^2 + \epsilon] \leq e^{-nJ_{LAG}(\epsilon)}$$

Modified AIC



Example:



Detection of Signals - Theoretical Questions

- ▶ Which Test Statistic to use ?
- ▶ Why does AIC overestimate number of signals ?
- ▶ **Detection Performance ?**
- ▶ Known vs. unknown noise level
- ▶ Non-parametric vs fully parametric setting.

Detection Performance

Depends on behavior of largest signal eigenvalue in presence of noise. With p fixed, $n \rightarrow \infty$,

$$P_d^{Roy} \approx Q \left[\sqrt{n} \left(\frac{th(\alpha)}{1 + \lambda/\sigma^2} - \frac{p-1}{N\lambda/\sigma^2} - 1 \right) \right]$$

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$$P_d^{GLRT} \approx Q \left[\sqrt{n} \left(\tilde{th}(\alpha) \left(\frac{1}{1 + \lambda/\sigma^2} - \frac{1}{n\lambda/\sigma^2} \right) - \frac{p-1}{N\lambda/\sigma^2} - 1 \right) \right]$$

where $Q(z) = \frac{1}{\sqrt{2\pi}} \int_z^\infty e^{-x^2/2} dx$,

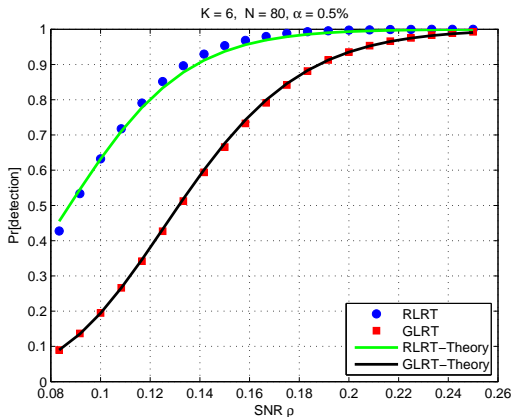
$$\tilde{th}(\alpha) = \frac{p-1}{p - th_U(\alpha)} th_U(\alpha) > th(\alpha)$$

[Kritchman & N. 09']

[N, Penna, Garello, submitted]

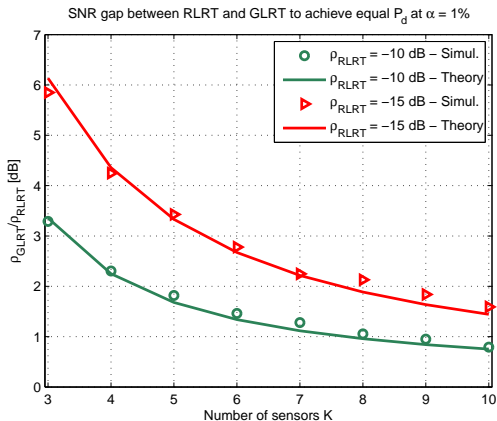
Known vs. Unknown Noise Level

Difference can be large:



Known vs. Unknown Noise Level

Difference can be large (several dB)



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Non-Parametric vs. Parametric Detection

In signal array processing, the array manifold is typically known.
For Uniform Linear Array

$$\mathbf{a}(\theta) = \left[1 e^{i\pi \sin \theta} e^{2i\pi \sin(\theta)} \dots e^{(p-1)i\pi \sin \theta} \right]$$

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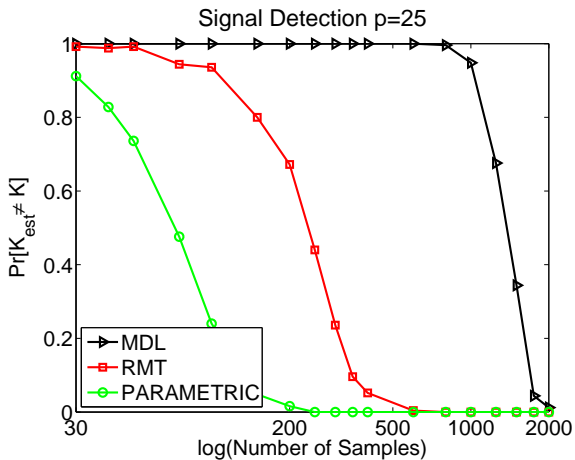
Theorem: Parametric Signal Detection requires

$$SNR \gg \sqrt{\frac{2 \ln p}{n}}$$

instead of

$$SNR \gg \sqrt{\frac{p}{n}}$$

Example:



Summary and Open Questions

Open Questions:

- Detection in unknown noise environments
- Non-Gaussian signals
- Sparse covariance matrices

<http://www.wisdom.weizmann.ac.il/~nadler/>

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C'est Tou / Merci Beaucoup