

Capacity of the random Code-Division-Multiple-Access channel with binary inputs

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Joint work with Satish Korada (Stanford).

References:

- ▶ Proc ISIT (2006)
- ▶ Proc 45-th Allerton Conf Communication, Control, Computing (2007)
- ▶ arXiv:0803.1454 (cs.IT); submitted to IEEE Trans Inf Theory

PLAN

- ▶ Code-Division-Multiple-Access setting
- ▶ Results in literature for gaussian and binary inputs
- ▶ **New contributions for binary inputs**
- ▶ Statistical mechanics formulation
- ▶ Interpolation method

DEFINITION OF CDMA

Users $k = 1, \dots, K$ transmit $x_k(1)x_k(2)x_k(3)\dots$ over a common gaussian channel to a single receiver $y(1)y(2)y(3)\dots$

Users have N "degrees of freedom" available (time slots, frequencies, ...) $j = 1, \dots, N$.

- ▶ **Code division:** user k "spreads" its current symbol x_k over the N "degrees of freedom" and transmits the vector

$$\frac{1}{\sqrt{N}} \mathbf{s}_{kj} x_k, \quad \mathbb{E}[x_k^2] = 1, \quad \frac{1}{N} \sum_{j=1}^N s_{kj}^2 = 1$$

- ▶ **Received signal:** vector $j = 1, \dots, N$

$$y_j = \frac{1}{\sqrt{N}} \sum_{k=1}^K \mathbf{s}_{kj} x_k + \sigma n_j, \quad \mathbb{E}[n_j^2] = 1$$

- ▶ **Conventional CDMA:** (*Verdu 1986*) spreading sequences are fixed.

$$C = \max_{\prod p_{x_k}} \frac{1}{K} I(\underline{X}; \underline{Y}) = \frac{1}{2K} \log \det(I_K + \sigma^{-2} \mathbf{S} \mathbf{S}^t)$$

The max is attained at standard **gaussian distribution for inputs**.

- ▶ **Random model:** (*Verdu-Shamai 1999*) spreading sequence i.i.d standard gaussian s_{kj} ; look at $\lim_{K \rightarrow \infty}$ with $\frac{K}{N} = \beta$ fixed.

$$C = \underbrace{\frac{1}{2K} \mathbb{E} \mathbf{s} \log \det(I_K + \sigma^{-2} \mathbf{S} \mathbf{S}^t)}_{\text{can be calculated by RMT}}$$

- ▶ Discrete input alphabets: $p_{X_k} = p_k \delta_{-1} + (1 - p_k) \delta_{+1}$

$$\max_{\prod p_{X_k}} \frac{1}{K} I(\underline{X}; \underline{Y})$$

is not known.

- ▶ Random model with discrete inputs: i.i.d standard gaussian s_{kj} ; look at $\lim_{K \rightarrow \infty}$ with $\frac{K}{N} = \beta$ fixed.

$$C = \underbrace{\mathbb{E}_{\mathbf{s}} \frac{1}{K} I(\underline{X}; \underline{Y})}_{\text{no logdet RMT formula}}, \quad p_k = \frac{1}{2}$$

no logdet RMT formula

TANAKA'S FORMULA (2001)

Using the formal "replica method" of statistical mechanics one reduces the problem to a variational problem,

$$\lim_{K \rightarrow \infty} C = \min_{m \in [0,1]} c(m), \quad \beta = \frac{K}{N} \text{ fixed}$$

For binary input symbols $x_k \in \{+1, -1\}$ and any symmetric distr for s_{kj} with finite second and fourth moments:

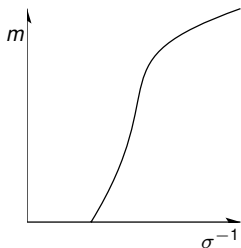
$$c(m) = \frac{\lambda}{2}(1+m) - \frac{1}{2\beta} \ln \lambda \sigma^2 - \int dz \frac{e^{-\frac{z^2}{2}}}{\sqrt{2\pi}} \ln(2 \cosh(\sqrt{\lambda}z + \lambda))$$

with

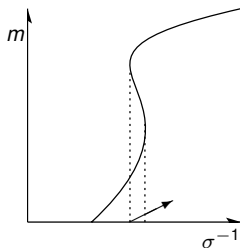
$$\lambda = \frac{1}{\sigma^2 + \beta(1-m)}$$

The minimizer m_* is one of the solutions of

$$m = \int dz \frac{e^{-\frac{z^2}{2}}}{\sqrt{2\pi}} \tanh(\sqrt{\lambda}z + \lambda), \quad \lambda = \frac{1}{\sigma^2 + \beta(1 - m)}$$



$(\beta < \beta_u)$



$(\beta > \beta_u)$

- ▶ $\beta < \beta_u$ unique solution: $m_*(\sigma)$ continuous.
- ▶ $\beta > \beta_u$ many solutions: $m_*(\sigma)$ first order phase transition.

The replica method is very powerful...

- ▶ "Any" type of input symbol: discrete or continuous. For gaussian inputs,

$$c(m) = \frac{1}{2} \ln(1 + \lambda) - \frac{1}{2\beta} \ln \lambda \sigma^2 - \frac{\lambda}{2} (1 - m)$$

$C = \min_{m \in [0,1]} c(m)$ agrees with RMT.

- ▶ Unequal powers for users.
- ▶ Colored noise
- ▶ Communication on CDMA channel with LDPC codes.

(Tanaka, Guo-Verdu, Kabashima-Saad, ...)

RIGOROUS CONTRIBUTIONS (binary inputs)

General assumption: i.i.d

$$p(s_{kj}) = p(-s_{kj}), \quad p(s_{kj} \geq s) \leq e^{-As^2} \text{ if } s \geq s_0$$

Theorem (S. Korada, N.M 2007)

- ▶ *The $\lim_{K \rightarrow \infty} C$ exists and is equal to $\lim_{K \rightarrow \infty} C_g$ where C_g is the capacity for gaussian $p(s_{kj})$.*
- ▶ *Tanaka's formula is a lower bound for all β*

$$\lim_{K \rightarrow \infty} C \leq \min_{m \in [0,1]} c(m)$$

Montanari and Tse (ITW 2005) sketch the derivation of a lower bound on,

$$\lim_{K \rightarrow \infty} \frac{d}{d\sigma} C, \quad \text{all } \beta$$

- ▶ For $\beta \leq \beta_u$ there is no phase transition and by integrating properly the bound you get $\lim_{K \rightarrow \infty} C_K = \min_{m \in [0,1]} c(m)$ for all σ .
- ▶ For $\beta \geq \beta_u$ there is a phase transition at $\sigma_c(\beta)$. Above the critical noise their bound is the same than ours. *Well below* the critical noise their bound is the converse one so by combining their result with ours one gets again the equality.

STATISTICAL MECHANICS FORMULATION

To compute $C = \ln 2 - \frac{1}{K} \mathbb{E}_{\mathbf{s}} H(\underline{X} | \underline{Y})$ we consider the posterior

$$p(\underline{x} | \underline{y}, \mathbf{s}) = \frac{1}{Z(\underline{y}, \mathbf{s})} e^{-\frac{1}{2\sigma^2} \|\underline{y} - N^{-\frac{1}{2}} \mathbf{s} \underline{x}\|^2}$$

with

$$Z(\underline{y}, \mathbf{s}) = \sum_{\underline{x}} e^{-\frac{1}{2\sigma^2} \|\underline{y} - N^{-\frac{1}{2}} \mathbf{s} \underline{x}\|^2}$$

and

$$p(\underline{y} | \mathbf{s}) = \sum_{\underline{x}^{input}} \frac{1}{2^K} \frac{e^{-\frac{1}{2\sigma^2} \|\underline{y} - N^{-\frac{1}{2}} \mathbf{s} \underline{x}^{input}\|^2}}{(\sqrt{2\pi}\sigma^2)^N}$$

and

$$p(\mathbf{s}) \quad i.i.d \quad gaussian$$

This leads to

$$C_K = \ln 2 - \frac{1}{2\beta} - \frac{1}{K} \mathbb{E}_{\underline{y}, \mathbf{s}} [\ln Z(\underline{y}, \mathbf{s})]$$

Fundamental object of stat mech "free energy"

$$\frac{1}{K} \ln Z(\underline{y}, \mathbf{s})$$

where Z is the "partition function"

$$Z(\underline{y}, \mathbf{s}) = \sum_{\underline{x}} e^{-\frac{1}{2\sigma^2} \|\underline{y} - N^{-\frac{1}{2}} \mathbf{s} \underline{x}\|^2} = \sum_{\underline{x}} e^{-\mathcal{H}(\underline{x})}$$

and $\mathcal{H}(\underline{x})$ is the "Hamiltonian" or cost function (log of channel transition probability).

For CDMA the Hamiltonian is

$$\mathcal{H}(\underline{x}) = \frac{\sqrt{\beta}}{2\sigma^2\sqrt{K}} \sum_{k,l} J_{kl} x_k x_l - \frac{1}{\sigma^2} \sum_{k=1}^K h_k x_k + \frac{1}{2\sigma^2} \|\underline{y}\|^2$$

with

$$J_{kl} = \frac{1}{\sqrt{N}} \sum_{j=1}^N s_{kj} s_{lj}, \quad h_k = \frac{1}{\sqrt{N}} \sum_{j=1}^N y_j s_{kj}$$

- ▶ Spins $x_k \in \{-1, +1\}$ are the dynamical degrees of freedom.
- ▶ Couplings J_{kl}, h_k are frozen/quenched disorder.

- ▶ CDMA is a complicated **spin glass model**.
- ▶ Superficially similar to the Sherington-Kirkpatrick model

$$\mathcal{H}(\underline{x}) = \frac{1}{\sqrt{K}} \sum_{k,l} J_{kl} x_k x_l \quad \text{i.i.d } J_{kl} \text{ distr } \mathcal{N}(0, J)$$

- ▶ If you change $\mathcal{H}(\underline{x}) \rightarrow -\mathcal{H}(\underline{x})$ you get a kind of Hopfield Hamiltonian.
- ▶ As for other communications problems: **Nishimori gauge symmetry** \rightarrow replica symmetric solution is expected to be correct.

THE INTERPOLATION METHOD

It was pioneered by Guerra-Toninelli. Based on it Talagrand arrived at a proof of the Parisi formula for SK.

- ▶ Takes the replica solution as the favorite guess and tries to find the corresponding "mean field Hamiltonian or channel"
- ▶ Construct an interpolating Hamiltonian or channel:
 $0 \leq t \leq 1$
- ▶ Fundamental theorem of calculus

$$\underbrace{\ln Z(1)}_{\text{true system}} = \underbrace{\ln Z(0)}_{\text{mean field syst}} + \int_0^1 dt \frac{d}{dt} \underbrace{\ln Z(t)}_{\text{interpolating syst}}$$

- ▶ The derivative produces correlation functions with a controllable sign (hopefully).

Step 1. Guessing: $\lim_{K \rightarrow \infty} C = \min_{m \in [0,1]} c(m)$

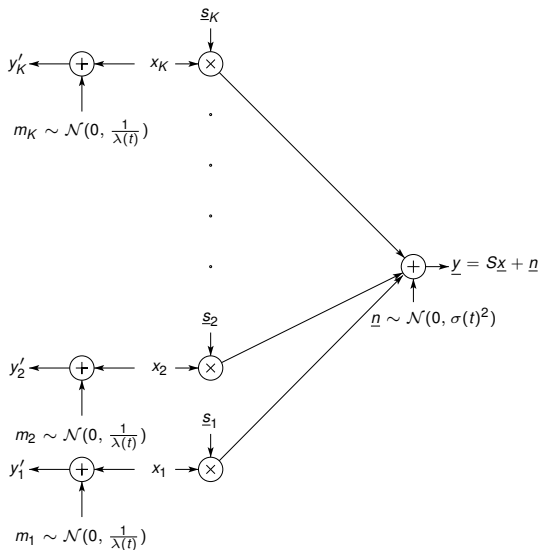
$$c(m) = \frac{\lambda}{2}(1+m) - \frac{1}{2\beta} \ln \lambda \sigma^2 - \underbrace{\int dz \frac{e^{-\frac{z^2}{2}}}{\sqrt{2\pi}} \ln(2 \cosh(\sqrt{\lambda}z + \lambda))}_{\text{almost capacity of BIAWGN}(\lambda^{-1})}$$

Mean field Hamiltonian correspond to K independent BIAWGN channels

$$y'_k = x_k + \lambda^{-1/2} m_k, \quad m_k \sim \mathcal{N}(0, \lambda^{-1})$$

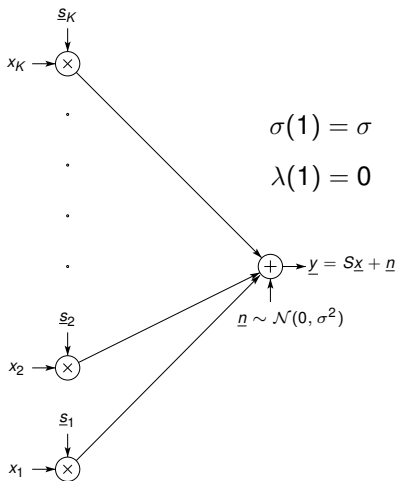
Recall $\lambda^{-1} = \sigma^2 + \beta(1-m)$.

Step 2. Interpolation channel $0 \leq t \leq 1$:

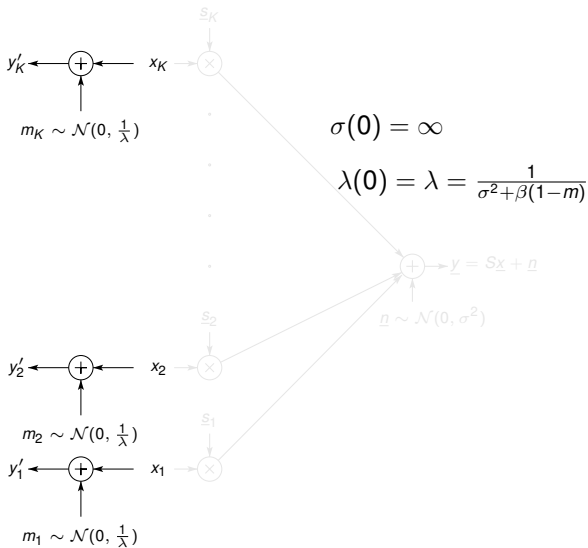


$$\lambda(t) + \frac{1}{\sigma^2(t) + \beta(1 - m)} = \frac{1}{\sigma^2 + \beta(1 - m)}$$

t=1 is the original CDMA channel



$t=0$ are K independent BIAWGN channels



Capacity of interpolating system

$$\frac{1}{K} \mathbb{E}_{\mathbf{s}} I_t(\underline{X}; \underline{Y}, \underline{Y}') = \ln 2 - \frac{1}{2\beta} - \mathbb{E}_{\mathbf{s}, \underline{Y}, \underline{Y}'} \ln Z(t)$$

where

$$Z(t) = \sum_{\underline{x}} e^{-\frac{1}{2\sigma(t)^2} \|\underline{y} - N^{-1/2} \mathbf{S} \underline{x}\|^2 - \frac{\lambda(t)}{2} \|\underline{y}' - \underline{x}\|^2}$$

and

$$\underline{y} = \frac{1}{\sqrt{N}} \mathbf{S} \underline{x}^{input} + \underline{n}, \quad n_i \sim \mathcal{N}(0, \sigma(t)^2)$$

$$\underline{y}' = \underline{x}^{input} + \underline{m}, \quad m_k \sim \mathcal{N}(0, \frac{1}{\lambda(t)}).$$

Step 3. Fundamental theorem of calculus:

$$\mathbb{E}_{\mathbf{s}, \underline{Y}, \underline{Y}'} \ln Z(1) = \mathbb{E}_{\mathbf{s}, \underline{Y}, \underline{Y}'} \ln Z(0) + \int_0^1 dt \frac{d}{dt} \mathbb{E}_{\mathbf{s}, \underline{Y}, \underline{Y}'} \ln Z(t)$$

This leads to

$$C = c(m) + \int_0^1 dt R(t) + o_K(1)$$

with

$$R(t) \leq 0, \quad \text{for all } 0 \leq t \leq 1$$

Step 4. Sign of remainder: (hard to bring in ratio form)

$$R(t) = -\beta \frac{\sigma'(t)}{\sigma(t)} \frac{(\mathbb{E}_{\underline{y}, \underline{y}', \mathbf{s}} \langle \mu - m \rangle_t)^2}{(\sigma(t)^2 + \beta(1 - m))(\sigma(t)^2 + \beta \mathbb{E}_{\underline{y}, \underline{y}', \mathbf{s}} \langle 1 - \mu \rangle_t)}$$

where

$$\mu = \frac{1}{K} \sum_{k=1}^K x_k^{\text{input}} x_k \quad (\text{magnetization})$$

$$\langle - \rangle_t \text{ is } p_t(\underline{x} | \underline{y}, \underline{y}', \mathbf{s}) = \frac{e^{-\mathcal{H}_t(\underline{x})}}{Z_t} \quad (\text{interpolating Gibbs})$$

Main ingredients for calculating the remainder

$$\frac{d}{dt} \mathbb{E}_{\underline{Y}, \underline{Y}', \mathbf{s}} \ln Z(t) = \mathbb{E}_{\underline{Y}, \underline{Y}', \mathbf{s}} \langle \text{horrible polynomial}(\underline{x}, \underline{y}, \underline{y}', \mathbf{s}) \rangle_t$$

- ▶ Integration by parts formula for Gaussian r.v (Guerra)

$$\mathbb{E}[u\varphi(u)] = \mathbb{E}[\varphi'(u)]$$

- ▶ Gauge symmetry
- ▶ Concentration theorems

Gauge symmetry implies remarkable identities (Nishimori)

$$\mu = \frac{1}{K} \sum_{k=1}^K x_k^{\text{input}} x_k \quad \text{and} \quad q = \frac{1}{K} \sum_{k=1}^K x_k^{(1)} x_k^{(2)}$$

have the same distribution

$$\mathbb{E}_{\underline{Y}, \underline{Y}', \mathbf{s}} \langle \mu^p \rangle_t = \mathbb{E}_{\underline{Y}, \underline{Y}', \mathbf{s}} \langle q^p \rangle_t$$

- ▶ Such identities arise in a natural way for various channel models.

Concentration: we need $\mathbb{E}\langle(\mu - \mathbb{E}\langle\mu\rangle)^2\rangle \rightarrow 0$

Applying concentration thms for Lipschitz functions of i.i.d. gaussians:

Theorem

For $p(s_{kj})$ standard gaussian:

$$\mathbb{P}[|\ln Z(\underline{y}, \mathbf{s}) - \mathbb{E}_{\underline{y}, \mathbf{s}} \ln Z(\underline{y}, \mathbf{s})| \geq \epsilon K] \leq e^{-\alpha(\beta, \sigma) \epsilon^2 \sqrt{K}}$$

$$\mathbb{P}[|I(\underline{X}; \underline{Y}) - \mathbb{E}_{\mathbf{s}} I(\underline{X}; \underline{Y})| \geq \epsilon K] \leq e^{-\alpha(\beta, \sigma) \epsilon^2 K}$$

Remark that

$$\langle \mu \rangle = \frac{\partial}{\partial u} \ln \underbrace{\sum_{\underline{x}} e^{-\mathcal{H}(\underline{x}) + u \sum_k x_k^{\text{input}} x_k}}_{Z_u}$$

Turn concentration of $\ln Z_u$ into

Corollary

Fix $\epsilon > 0$. For Lebesgue almost every $u > \epsilon$,

$$\lim_{K \rightarrow \infty} \int_0^1 dt \mathbb{E} \langle (\mu - \mathbb{E} \langle \mu \rangle_{t,u})^2 \rangle_{t,u} = 0$$

In reality the u -perturbation is more subtle because our methods do not afford to break the channel symmetry.

CONCLUSION

- ▶ Other sorts of interpolations allow to prove that $\lim C_K$ exists and is independent of distribution of spreading sequence.
- ▶ More general input distributions, Unequal powers of users, MIMO , CDMA with users using LDPC Codes, Coloured noise.
- ▶ Main open question: lower bound on Capacity. We can do this for simpler special spin glasses with gauge symmetry.
- ▶ For LDPC codes over BMS channels the interpolation method has been developed but the lower bound is also missing.